

# Lesson 3: Fading Memory Nonlinearities

### Nonlinear Signal Processing - SS 2019

Christian Knoll Signal Processing and Speech Communication Laboratory Graz University of Technology

May 23, 2019

**NLSP SS 2019** 

May 23, 2019

∃ ∽ Slide 1/17

イロン イヨン イヨン イヨン



### Session contents

- Today:
  - Volterra Series as representation of fading-memory NL
  - System identification using different inputs
  - Time series modelling (Homework)
- Next time:
  - Higher-order statistics and spectral analysis



### Volterra series – Definition

► A finite Volterra series of order *p* and memory length *M*:

$$y[n] = h_0 + \sum_{m_1=0}^{M-1} h_1[m_1] \times [n - m_1]$$
  
+ 
$$\sum_{m_1=0}^{M-1} \sum_{m_2=0}^{M-1} h_2[m_1, m_2] \times [n - m_1] \times [n - m_2] +$$
  
+ 
$$\sum_{m_1=0}^{M-1} \sum_{m_2=0}^{M-1} \sum_{m_3=0}^{M-1} h_3[m_1, m_2, m_3] \times [n - m_1] \times [n - m_2] \times [n - m_3] + \dots +$$
  
+ 
$$\sum_{m_1=0}^{M-1} \cdots \sum_{m_p=0}^{M-1} h_p[m_1, \dots, m_p] \prod_{i=1}^{p} \times [n - m_i]$$

 Universal approximator for time-invariant causal operators with fading memory (for bounded input signals)

**NLSP SS 2019** 

(日) (四) (日) (日) (日)



# Volterra series – System identification (1)

- ► Only term with h<sub>1</sub>[·] is linear!
- Complexity grows exponentially with M<sup>p</sup>
- Choice of input signal for system identification?
- $\blacktriangleright$  Remember RBF-fits: Model also linear in coefficients  $\rightarrow$  Least-squares fit easy

$$y[n] = \sum_{k} \alpha_k \phi_k(x[n])$$
 N equations

Arrange in equation system

$$\underbrace{\begin{bmatrix} y[0] \\ \vdots \\ y[N-1] \end{bmatrix}}_{\mathbf{y}} = \underbrace{\begin{bmatrix} \phi_1(x[0]) & \cdots & \phi_K(x[0]) \\ \vdots & \ddots & \vdots \\ \phi_1(x[N-1]) & \cdots & \phi_K(x[N-1]) \end{bmatrix}}_{\Phi, \ (N \times K), \ N \gg K} \begin{bmatrix} \alpha_1 \\ \vdots \\ \alpha_K \end{bmatrix}$$

イロン イヨン イヨン イヨン



# Volterra series – System identification (2)

- Here we use the same method, just different basis functions
- Second order Volterra system as example  $(1 + M + M^2 \text{ coeff.})$
- You need "quite some" data!

$$y[n] = h_0 + \sum_{m_1=0}^{M-1} h_1[m_1] \times [n-m_1] + \sum_{m_1=0}^{M-1} \sum_{m_2=0}^{M-1} h_2[m_1, m_2] \times [n-m_1] \times [n-m_2]$$

$$\begin{bmatrix} y[0] \\ y[1] \\ \vdots \\ y[N-1] \end{bmatrix} = \begin{bmatrix} 1 & x[0] & \cdots & x[-M+1] & x^2[0] & x[0]x[-1] & \cdots & x^2[-M+1, -M+1] \\ 1 & x[1] & \cdots & x[-M+2] & x^2[1] & x[1]x[0] & \cdots & x^2[-M+2, -M+2] \\ \vdots & & \vdots \end{bmatrix} \begin{bmatrix} h_0 \\ h_1[0] \\ \vdots \\ h_1[M-1] \\ h_2[0, 0] \\ h_2[0, 1] \\ \vdots \\ h_2[M-1, M-1] \end{bmatrix}$$

### Basis vectors/functions are the data products

NLSP SS 2019	May 23, 2019	Slide 5/17

オロト オポト オモト オモト・モ



### Volterra series – Matlab files

- Function vkernels.m does the LS-fit (use p3\_1.m as tutorial)
- Nonlinearities defined in nlsystem1.m

$$y[n] = a_0 x[n] + a_1 x[n-1] + a_2 x[n]^2 + a_3 x[n] x[n-1]$$

Output of vkernels.m for this nonlinearity:

$$\begin{aligned} & \operatorname{H}\{1\} = 0 \\ & \operatorname{H}\{2\} = \begin{bmatrix} a_0 & a_1 \end{bmatrix}^T \\ & \operatorname{H}\{3\} = \begin{bmatrix} x[n]^2 & x[n]x[n-1] \\ & a_2 & a_3 \\ & a_3 & 0 \\ & x[n-1]x[n] & x[n-1]^2 \end{bmatrix} \end{aligned}$$

Or optionally a structure Vmodel that can be directly passed to vkernels\_o.m for Problem 3.2

■ 
Slide 6/17



# Volterra series – Time series modelling/forecasting

- ► Given a correlated (not necessarily just second order!) time series s = [s<sub>1</sub>,..., s<sub>N</sub>]<sup>T</sup>
- Current sample s<sub>n</sub> depends on past samples
- Volterra series one way to model the dependence
- Input x and output y both generated from s
- Partitioning of s into training and validation sequences
- $\rightarrow$  Homework

(日) (同) (E) (E) (E)



### Higher order statistics and spectral analysis

We are used to first- and second-order statistics

• Mean: 
$$\mu = \mathsf{E}\{x[n]\}$$

• ACF:  $m_2[k] = E\{x[n]x[n+k]\}$ 

This is suitable as long as we deal with linear systems, e.g.

$$y[n] = \sum_{k=0}^{K} x[k]h[n-k]$$

- Standard example for a random process: Linear process, i.e.
   x[k] is a white-noise process driving a linear system
- But what if our system is nonlinear, e.g.

$$y[n] = a_0 x[n] + a_1 x[n-1] + a_2 x[n]^2 + a_3 x[n] x[n-1]$$

・ロ・ ・ 日・ ・ 日・ ・ 日・



• Linear system w. K = 10, driven by white noise



イロン イヨン イヨン イヨン



• Linear system w. K = 10, driven by white noise



Support of non-zero ACF indication for memory

イロン イヨン イヨン イヨン



• Linear system w. K = 10, driven by white noise



- Support of non-zero ACF indication for memory
- How do you evaluate nonlinear combinations of input samples?

(日) (四) (日) (日) (日)



• Linear system w. K = 10, driven by white noise



- Support of non-zero ACF indication for memory
- How do you evaluate nonlinear combinations of input samples?
- $\rightarrow$  Evaluate HOS, e.g. third-order E{x[n]x[n + k]x[n + I]}

イロト イヨト イヨト イヨト



# HOSA – Definitions

 Generalization of ACF: Non-central moments of order r for stationary process x[n]

$$m_r[k_1,\ldots,k_{r-1}] = \mathsf{E}\{x[n]x[n+k_1]\cdots x[n+k_{r-1}]\}$$

- Cumulant of order r again defined over characteristic function
- Can be expressed and estimated via moments
- Fourier transform of ACF is power spectral density
- Fourier transform (2D) of third order cumulant is the Bispectrum

$$\triangleright \ \mathsf{c}_3^{\mathsf{x}}[k_1, k_2] \quad \stackrel{\mathsf{DFT}}{\longrightarrow} \quad \mathcal{C}_3^{\mathsf{x}}[\omega_1, \omega_2]$$

ヘロト 人間 とくほと くほとう



### HOSA – Important Properties

x[n] Gaussian:  $c_r^{x}(k_1,\ldots,k_{r-1}) = C_r^{x}(\omega_1,\ldots,\omega_{r-1}) = 0$  for r > 2▶ x[n] i.i.d.:  $c_{r}^{x}(k_{1},\ldots,k_{r-1}) = a \cdot \delta(k_{1},\ldots,k_{r-1})$  $\mathcal{C}_{r}^{x}(\omega_{1},\ldots,\omega_{r-1})=a$ x[n] symmetrically distributed around zero:  $c_r^{x}(k_1,\ldots,k_{r-1}) = C_r^{x}(\omega_1,\ldots,\omega_{r-1}) = 0$  for  $r = 0, 3, 5, 7, \ldots$ rightarrow z[n] = x[n] + y[n], where x[n], y[n] jointly stationary and statistically independent:  $c^{z}(x) = c^{x}(x) \pm c^{y}(x)$ 

$$\mathcal{C}_r^z(\cdot) = \mathcal{C}_r^z(\cdot) + \mathcal{C}_r^y(\cdot)$$
$$\mathcal{C}_r^z(\cdot) = \mathcal{C}_r^x(\cdot) + \mathcal{C}_r^y(\cdot)$$

**NLSP SS 2019** 

Slide 11/17

◆□> ◆□> ◆三> ◆三> ●三 のへの



### HOSA – Pros and Cons

### Pros

- Analysis of nonlinearities
- Cumulants are additive for independent processes
- Gaussian noise: HOS zero (blind to Gaussian noise)

### Cons

- Difficult to estimate from finite length data
- Influence of window
- Once you have them, how do you interpret them?



### HOSA – Example

### Example for a third-order cumulant



**NLSP SS 2019** 

May 23, 2019

≣ • ? Q ( Slide 13/17



### HOSA – Example

### Example for a Bispectrum



**NLSP SS 2019** 

May 23, 2019

≣ ∽ ۹ Slide 13/17



### HOSA – Example

Example for a PSD





≣ ∽ ۹ Slide 13/17



# HOSA – Matlab (1)

- HOSA toolbox, free, included in download-file
- ▶ HOSA toolbox manual is a great ressource (Matlab-central)!
- You will need:
  - cumest.m used to estimate cumulants
  - rpiid.m used to help generating input processes
  - gabrrao.m used to calculate window for 2D-FFT
  - viscumul3.m and
  - visbispec3.m for visualization



# HOSA – Matlab (2)

 cumest.m for third order cumulant calculates just one slice of the 2D correlation function

- Bispectrum calculation: use fftshift(fft2( c3 .\* w )) to have a familiar picture
- Window w[n] obtained from gabrrao.m, optimal smoothing window, minimum bias in estimation

◆□▶ ◆□▶ ◆三▶ ◆三▶ ○○ ○○



- Limited amount of data leads to higher variance of estimators
- $\blacktriangleright$  Problem even for ACF  $\rightarrow$  Grows exponentially with order
- This makes visual interpretation much harder:
  - When is a Bispectrum zero?
  - When is a Bicoherence function flat?



• Linear system w. K = 10, driven by white noise (500 samples)



ж

・ロン ・回 と ・ ヨン・



• Linear system w. K = 10, driven by white noise (500 samples)



**NLSP SS 2019** 

May 23, 2019

Slide 17/17

(日) (四) (日) (日) (日)



• Linear system w. K = 10, driven by white noise (500 samples)



**NLSP SS 2019** 

Slide 17/17

э

・ロン ・回 と ・ ヨン・



• Linear system w. K = 10, driven by white noise (500 samples)



**NLSP SS 2019** 

Slide 17/17

э

・ロン ・回 と ・ ヨン・